

Fig. 1. Binary bit error rates of NCOOK and CPSK systems operating under atmospheric burst noise interference. Dotted curve gives variation of threshold constant α with SNR for NCOOK systems.

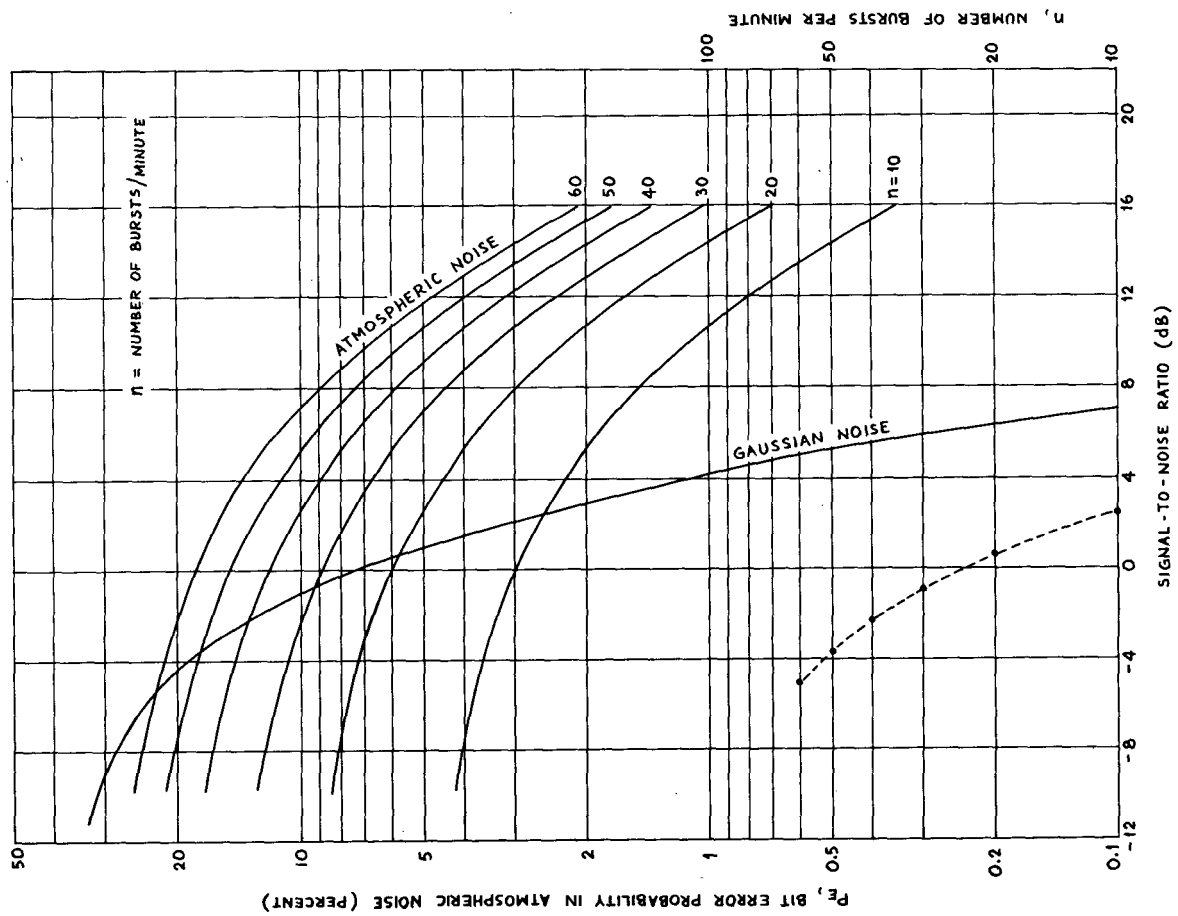


Fig. 2. Comparison of binary bit error rates of CPSK system in Gaussian noise and in atmospheric burst noise at various burst rates. Dotted curve relates α and corresponding SNR for which α_g is greater for atmospheric noise than for Gaussian noise.

called a noise burst. The noise burst is composed of distinct pulses of varying amplitude with duration of a pulse being determined by the bandwidth of the receiver. The number of pulses in a specified range of level from the peak pulse amplitude increases with frequency [2], [3] in the frequency range in which atmospheric noise is considered the principal source of interference, i.e., up to 30 MHz.

B. Short-Term Time Parameters of Noise Bursts [4], [5].

The duration of a noise burst changes randomly. Short-term studies on noise bursts arising from local thunderstorms show that the duration t_d is log-normally distributed with a median $t_{d,m}$ of 500 ms and standard deviation of 0.63 natural log units. Short-term studies on noise burst rate (number of bursts per minute) n revealed that the burst repetition period t_r is also log-normally distributed with median $t_{r,m}$ of $60/n$ s and standard deviation of 0.44 natural log units. Long-term data on n is available for tropical regions.

III. ESTIMATION OF BIT ERROR RATES

Consider the situation in which atmospheric noise is the principal source of interference and assume that errors can occur in the reception only during the presence of noise bursts. The noise bursts occur in a sequence with random intervals between the bursts. At the transmission speeds that are commonly in practice, several signal symbols occur within the duration of a burst. All the signal symbols (each signal symbol may be represented by several cycles of modulated sinusoid) will not be corrupted to the same extent due to the random variation in the amplitudes of the pulses and random occurrences of pulses themselves within the burst. Due to the large number of signal symbols that occur within the duration of the burst (150–1500 symbols for a 1000-baud transmission), a statistical investigation can be attempted to yield a probability of binary bit error p_e of receiving a binary symbol in error within the burst. A prerequisite for such an analysis is the amplitude distribution $p(N)$ of the noise pulses within a burst at the operating frequency and bandwidth.

p_e for systems using different types of modulation of a sinusoidal carrier can be evaluated easily with the help of phasor diagrams, on the lines indicated by Montgomery [6] and Engel [7]. However, p_e does not fully represent the probability of error during the presence of a burst, since the burst is composed of distinct pulses and regions characterized by their absence. From the knowledge of the number of pulses in a burst, the time occupied by the pulses in the burst t_o can be calculated. If a reasonable assumption that the pulses are distributed uniformly in the burst is made, then the mean value of t_o/t_a gives probability p_p that each signal symbol meets with a noise pulse.

The overall bit error probability P_E in the presence of a sequence of distinct noise bursts is given by

$$P_E = p_e \cdot p_p \cdot \alpha$$

where α is the fraction of time occupied by the bursts in the total time of a long sequence, equal to t_d/t_r .

Data on long term variation of the burst occurrence rate n are available for tropical locations. p_p is to be evaluated from knowledge of the number of pulses in the noise bursts and is a function of operating frequency. p_p for HF bands tends to be nearly equal to unity since pulses occur fairly continuously throughout the bursts. However, in the lower frequency bands the pulses are distinct and widely separated, occupying insignificant portions of burst duration, and leading to extremely small values of p_p . Thus, under the same signal level conditions,

as the operating frequency is lowered, the increase in p_e due to higher noise levels is counteracted by the reduction in p_p .

Error rate calculations are carried out for noncoherent on-off keying (NCOOK) and coherent phase-shift keying (CPSK) systems on the lines indicated above, using noise data available for India [8].

Noise Data

Season: March–May—time-block, 20–24 h.
 Frequency: 1 MHz—bandwidth, 1 kHz/6 dB.
 Seasonal median noise level: 30 dB above 1 μ V/m.
 Short-term median noise level: 27 dB approximately.
 Short-term noise level distribution: log normal.
 Standard deviation: 8 dB.
 Short-term median peak amplitude: 32 dB.
 Short-term median rms amplitude: 29.5 dB.
 $p_p \approx 1, \quad n = 20.$

IV. CONCLUSIONS

The results are plotted in Fig. 1, in which SNR is the ratio of squares of rms signal level and median rms noise level. In the NCOOK case, the optimum value for threshold constant a for which P_E is minimum, is a function of SNR, and is also plotted in the same figure. a is greater than 1/2 even for SNR as high as 20 dB. Thus, the arbitrary choice of a as 1/2 for large SNR as was done in Montgomery's analysis [6] in the calculation of P_E versus SNR does not hold for atmospheric noise.

It is interesting to compare the error probabilities due to atmospheric noise with that of Gaussian noise. Fig. 2 gives the curves p_B versus SNR for CPSK signaling in atmospheric burst noise environment for different burst rate n and in Gaussian noise. Using these curves, another curve relating to n and corresponding SNR for which p_B for atmospheric noise (p_{BA}) is greater than p_B for Gaussian noise (p_{BG}) is plotted in the same figure. As n increases, SNR for which p_{BA} is greater than p_{BG} decreases. An important conclusion that can be drawn from this curve is that the interference from atmospheric noise is more severe than from Gaussian noise of the same rms level, for predetection SNR greater than about 2 dB.

The analysis described in this paper breaks down for the LF band since statistical analysis within the burst to arrive at p_e does not hold, due to the very small number of pulses observable in the bursts. In addition, since the dynamic range of the amplitudes of pulses is rather large in this frequency region, amplitude limiting occurs in the preliminary wide-band stages of the receiver. This limiting effect is to be taken into account to link the predetection SNR with receiver input SNR.

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